General Description

The MAX4249–MAX4257 low-noise, low-distortion operational amplifiers offer Rail-to-Rail[®] outputs and singlesupply operation down to 2.4V. They draw 400µA of quiescent supply current per amplifier while featuring ultra-low distortion (0.0002% THD), as well as low input voltage-noise density (7.9nV/ \sqrt{Hz}) and low input current-noise density (0.5fA/ \sqrt{Hz}). These features make the devices an ideal choice for portable/battery-powered applications that require low distortion and/or low noise.

For additional power conservation, the MAX4249/ MAX4251/MAX4253/MAX4256 offer a low-power shutdown mode that reduces supply current to 0.5µA and puts the amplifiers' outputs into a high-impedance state. The MAX4249-MAX4257's outputs swing rail-torail and their input common-mode voltage range includes ground. The MAX4250–MAX4254 are unitygain stable with a gain-bandwidth product of 3MHz. The MAX4249/MAX4255/MAX4256/MAX4257 are internally compensated for gains of 10V/V or greater with a gain-bandwidth product of 22MHz. The single MAX4250/MAX4255 are available in space-saving 5-pin SOT23 packages. The MAX4252 is available in an 8-pin ultra chip-scale package (UCSP™) and the MAX4253 is available in a 10-pin UCSP.

Applications

Wireless Communications Devices PA Control

Portable/Battery-Powered Equipment

Medical Instrumentation

ADC Buffers

Digital Scales/Strain Gauges

_Features

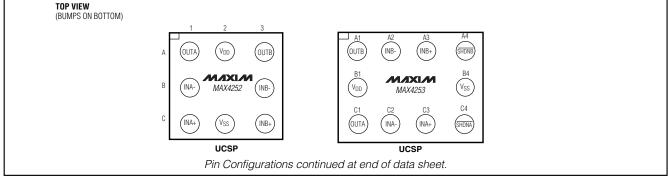
- Available in Space-Saving UCSP, SOT23, and µMAX Packages
- Low Distortion: 0.0002% THD (1kΩ load)
- 400µA Quiescent Supply Current per Amplifier
- Single-Supply Operation from 2.4V to 5.5V
- Input Common-Mode Voltage Range Includes Ground
- ♦ Outputs Swing Within 8mV of Rails with a 10kΩ Load
- ◆ 3MHz GBW Product, Unity-Gain Stable (MAX4250–MAX4254)
 22MHz GBW Product, Stable with A_V ≥ 10V/V (MAX4249/MAX4255/MAX4256/MAX4257)
- ♦ Excellent DC Characteristics
 V_{OS} = 70µV
 I_{BIAS} = 1pA
 Large-Signal Voltage Gain = 116dB
- Low-Power Shutdown Mode: Reduces Supply Current to 0.5µA Places Outputs in a High-Impedance State
- ♦ 400pF Capacitive-Load Handling Capability

_Ordering Information

PART	TEMP RANGE	PIN- PACKAGE	TOP MARK
MAX4249ESD	-40°C to +85°C	14 SO	_
MAX4249EUB	-40°C to +85°C	10 µMAX	—
MAX4250EUK-T	-40°C to +85°C	5 SOT23-5	ACCI

Ordering Information continued at end of data sheet. Selector Guide appears at end of data sheet.

Pin Configurations



Rail-to-Rail is a registered trademark of Nippon Motorola, Ltd. UCSP is a trademark of Maxim Integrated Products, Inc.

M/IXI/M

Maxim Integrated Products 1

For pricing, delivery, and ordering information, please contact Maxim/Dallas Direct! at 1-888-629-4642, or visit Maxim's website at www.maxim-ic.com.

ABSOLUTE MAXIMUM RATINGS

Power-Supply Voltage (VDD to VSS)	+6.0V to -0.3V
Analog Input Voltage (IN_+, IN)(VDD + (0.3V) to (V _{SS} - 0.3V)
SHDN Input Voltage	.6.0V to (V _{SS} - 0.3V)
Output Short-Circuit Duration to Either Supp	lyContinuous
Continuous Power Dissipation ($T_A = +70^{\circ}C$)	
5-Pin SOT23 (derate 7.1mW/°C above +7	70°C)571mW
8-Pin µMAX (derate 4.5mW/°C above +70	0°C)362mW
8-Pin SO (derate 5.88mW/°C above +70°	C)471mW
8-Pin UCSP (derate 4.7mW/°C above +70	0°C)379mW

10-Pin µMAX (derate 5.6mW/°C above +70°C)444mW 14-Pin SO (derate 8.33mW/°C above +70°C)667mW Operating Temperature Range40°C to +85°C	
Junction Temperature+150°C	
Storage Temperature Range65°C to +150°C	
Lead Temperature (soldering, 10s)+300°C	
Bump Temperature (soldering) (Note 1)	
Infrared (15s)+220°C	
Vapor Phase (60s)+215°C	

Note 1: This device is constructed using a unique set of packaging techniques that impose a limit on the thermal profile the device can be exposed to during board-level solder attach and rework. This limit permits only the use of the solder profiles recommended in the industry-standard specification, JEDEC 020A, paragrah 7.6, Table 3 for IR/VPR and Convection Reflow. Preheating is required. Hand or wave soldering is not allowed.

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

 $(V_{DD} = 5V, V_{SS} = 0, V_{CM} = 0, V_{OUT} = V_{DD}/2, R_L \text{ tied to } V_{DD}/2, \overline{SHDN} = V_{DD}, T_A = -40^{\circ}C \text{ to } +85^{\circ}C, \text{ unless otherwise noted. Typical values are at } T_A = +25^{\circ}C.)$ (Notes 2, 3)

PARAMETER	SYMBOL	с	CONDITIONS		MIN	ТҮР	МАХ	UNITS	
Supply Voltage Range	V _{DD}	(Note 4)			2.4		5.5	V	
			$V_{DD} = 3$	/		400			
Quiescent Supply Current Per	1-	Normal mode	$V_{DD} = 5$	/		420	575		
Amplifier	IQ		$V_{DD} = 5$, UCSP only		420	655	μA	
		Shutdown mode	e (SHDN =	V _{SS}) (Note 2)		0.5	1.5		
Input Offset Voltage (Note 5)	V _{OS}					±0.07	±0.75	mV	
Input Offset Voltage Tempco	TCVOS					0.3		µV/°C	
Input Bias Current	Ι _Β	(Note 6)				±1	±100	рА	
Input Offset Current	los	(Note 6)				±1	±100	рА	
Differential Input Resistance	R _{IN}					1000		GΩ	
Input Common-Mode Voltage Range	V _{CM}	Guaranteed by	CMRR test		-0.2		V _{DD} - 1.1	V	
Common-Mode Rejection Ratio	CMRR	$V_{SS} - 0.2V \le V_C$	$M \leq V_{DD} - T$	I.1V	70	115		dB	
Power-Supply Rejection Ratio	PSRR	V _{DD} = 2.4 to 5.5	ōV		75	100		dB	
	$V_{OUT} = 25 \text{mV} \text{ t}$				$V_{OUT} = 25 \text{mV}$ to $V_{DD} - 4.97 \text{V}$	80	116		٩D
Large-Signal Voltage Gain	Av	R_L = 1kΩ to V _{DD} /2; V _{OUT} = 150V to V _{DD} - 4.75V		80	112		dB		
Output Voltage Swing	Vout	$ V_{IN+} - V_{IN-} \ge 10$ $R_L = 10k\Omega$ to V_L		V _{DD} - V _{OH} V _{OL} - V _{SS}		8 7	25 20	mV	

ELECTRICAL CHARACTERISTICS (continued)

 $(V_{DD} = 5V, V_{SS} = 0, V_{CM} = 0, V_{OUT} = V_{DD}/2, R_L \text{ tied to } V_{DD}/2, \overline{SHDN} = V_{DD}, T_A = -40^{\circ}C \text{ to } +85^{\circ}C, \text{ unless otherwise noted. Typical values are at } T_A = +25^{\circ}C.)$ (Notes 2, 3)

PARAMETER	PARAMETER SYMBOL CONDITIONS		MIN	ТҮР	МАХ	UNITS		
Output Valtage Swing	Vour	IV _{IN+} - V _{IN-} I ≥ 10mV,	V _{DD} - V _{OH}		77	200	m)/	
Output Voltage Swing	Vout	$R_L = 1k\Omega$ to $V_{DD}/2$	V _{OL} - V _{SS}		47	100	mV	
Output Short-Circuit Current	ISC				68		mA	
Output Leakage Current	ILEAK	Shutdown mode (\overline{SHDN} = V_{OUT} = V_{SS} to V_{DD} (Note 2)			0.001	1.0	μA	
SHDN Logic Low	VIL	(Note 2)				0.2 x V _{DD}	V	
SHDN Logic High	VIH	(Note 2)		0.8 X V _{DD}			V	
SHDN Input Current	lıL/lıH	$\overline{\text{SHDN}} = V_{\text{SS}} = V_{\text{DD}}$ (Note	2)		0.5	1.5	μA	
Input Capacitance					11		pF	
	0.014/	MAX4250-MAX4254			3			
Gain-Bandwidth Product	GBW	MAX4249/MAX4255/MAX4	256/MAX4257		22		MHz	
	0.5	MAX4250-MAX4254		0.3		V/µs		
Slew Rate	SR	MAX4249/MAX4255/MAX4256/MAX4257			2.1	2.1		
Peak-to-Peak Input-Noise Voltage	e _{nP-P}	f = 0.1Hz to 10Hz		760			nV _{P-P}	
		f = 10Hz		2.7				
Input Voltage-Noise Density	en	f = 1kHz		8.9		nV/√Hz		
		f = 30kHz		7.9				
Input Current-Noise Density	in	f = 1kHz		0.5			fA/√Hz	
		MAX4250–MAX4254 Av = 1V/V, V _{OUT} = 2V _{P-P} ,	f = 1kHz		0.0004			
Total Harmonic Distortion Plus		$R_L = 1k\Omega$ to GND (Note 7)	f = 20kHz		0.006			
Noise	THD + N	MAX4249/MAX4255/ MAX4256/MAX4257	f = 1kHz		0.0012		%	
		$ A_V = 1V/V, V_{OUT} = 2V_{P-P}, \\ R_L = 1k\Omega \text{ to GND (Note 7)} $	f = 20kHz		0.007			
Capacitive-Load Stability		No sustained oscillations			400		pF	
		MAX4250–MAX4254, A _V = 1V/V			10			
Gain Margin	GM	MAX4249/MAX4255/MAX4256/MAX4257, A _V = 10V/V		12.5		dB		
		MAX4250-MAX4254, A _V =	1V/V		74			
Phase Margin	ФМ	MAX4249/MAX4255/MAX4 A _V = 10V/V	256/MAX4257,		68		degrees	

ELECTRICAL CHARACTERISTICS (continued)

 $(V_{DD} = 5V, V_{SS} = 0, V_{CM} = 0, V_{OUT} = V_{DD}/2, R_L \text{ tied to } V_{DD}/2, \overline{SHDN} = V_{DD}, T_A = -40^{\circ}C \text{ to } +85^{\circ}C, \text{ unless otherwise noted. Typical values are at } T_A = +25^{\circ}C.)$ (Notes 2, 3)

PARAMETER	SYMBOL	CONDITIONS		MIN	ТҮР	MAX	UNITS	
			MAX4250-MAX4254		6.7			
Settling Time		To 0.01%, VOUT = 2V step	MAX4249/MAX4255/ MAX4256/MAX4257		1.6		μs	
	n ^t SH n	t _{SH} IVDD = 5% of normal operation	MAX4251/MAX4253		0.8			
Delay Time to Shutdown			MAX4249/MAX4256		1.2		μs	
Delay Time to Enable	ten	VOUT = 2.5V, VOUT settles to	MAX4251/MAX4253		8		μs	
Doldy Hillo to Enable	-LIN	0.1%	MAX4249/MAX4256		3.5			
Power-Up Delay Time	tpu	$V_{DD} = 0$ to 5V ste	p, Vout stable to 0.1%		6		μs	

Note 2: SHDN is available on the MAX4249/MAX4251/MAX4253/MAX4256 only.

Note 3: All device specifications are 100% tested at $T_A = +25^{\circ}C$. Limits over temperature are guaranteed by design.

Note 4: Guaranteed by the PSRR test.

Note 5: Offset voltage prior to reflow on the UCSP.

Note 6: Guaranteed by design.

Note 7: Lowpass-filter bandwidth is 22kHz for f = 1kHz and 80kHz for f = 20kHz. Noise floor of test equipment = $10NV/\sqrt{Hz}$.

Typical Operating Characteristics

(V_{DD} = 5V, V_{SS} = 0, V_{CM} = V_{OUT} = V_{DD}/2, input noise floor of test equipment =10nV/√Hz for all distortion measurements, $T_A = +25^{\circ}C$, unless otherwise noted.) MAX4251/MAX4256 **OFFSET VOLTAGE INPUT OFFSET VOLTAGE vs. INPUT OFFSET VOLTAGE DISTRIBUTION** vs. TEMPERATURE **INPUT COMMON-MODE VOLTAGE** 40 250 200 400 UNITS $V_{CM} = 0$ 200 35 $V_{CM} = 0$ $T_A = +25^{\circ}C$ 150 150 INPUT OFFSET VOLTAGE (µV) 30 100 NUMBER OF UNITS 25 50 100 V0S (µV) 20 0 $V_{DD} = 5V$ -50 $V_{DD} = 3V$ 50 15 -100 10 -150 0 5 -200 0 -250 -50 -95 -75 -35 -13 -13 -40 -20 0 20 40 60 80 -0.5 0.5 1.5 2.5 3.5 4.5 TEMPERATURE (°C) $V_{0S}(\mu V)$ INPUT COMMON-MODE VOLTAGE (V) **OUTPUT VOLTAGE vs. OUTPUT VOLTAGE SWING (VOH) OUTPUT VOLTAGE SWING (VOL) OUTPUT LOAD CURRENT** vs. TEMPERATURE vs. TEMPERATURE 0.10 0.6 0.06 V_{DD} = 3V OR 5V 0.09 $V_{DIFF} = \pm 10 mV$ 0.5 0.05 0.08 $\dot{R}_{I} = 1k\Omega$ = 1kΩ VDD - VOH OUTPUT VOLTAGE (V) 0.07 0.4 0.04 VDD - VOH (V) 0.06 \geq 0.3 0.05 ੱਠ 0.03 Vni 0.04 0.2 0.02 0.03 0.02 0.1 0.01 $R_L = 10k\Omega$ $R_L = 10 k\Omega$ 0.01 $R_{I} = 100 k\Omega$ $R_{I} = 100 k\Omega$ 0 0 0 4 5 9 0 1 2 3 6 7 8 10 -40 -20 0 20 40 60 80 -40 -20 0 20 40 60 80 OUTPUT LOAD CURRENT (mA) TEMPERATURE (°C) TEMPERATURE (°C) LARGE-SIGNAL VOLTAGE GAIN LARGE-SIGNAL VOLTAGE GAIN LARGE-SIGNAL VOLTAGE GAIN **vs. OUTPUT VOLTAGE SWING vs. OUTPUT VOLTAGE SWING** vs. OUTPUT VOLTAGE SWING 140 140 140 = 200kΩ $B_1 = 200 k \Omega$ 130 130 130 $R_I = 20k\Omega$ 120 120 120 110 $R_L = 200 k\Omega$ 110 $R_L = 2k\Omega$ 110 = 20k Ω AV (dB) R AV (dB) $R_I = 20k\Omega$ AV (dB) 100 100 $R_L = 2k\Omega$ 100 90 90 90 $R_L = 2k\Omega$ 80 80 80 70 70 70 I V_{DD} = 3V R_L REFERENCED TO GND RL REFERENCED TO GND 60 R_I REFERENCED TO GND 60 50 60 50 50 100 150 200 250 0 50 100 150 200 250 0 0 50 250 100 150 200 VOUT SWING FROM EITHER SUPPLY (mV) V_{OUT} SWING FROM EITHER SUPPLY (mV) VOUT SWING FROM EITHER SUPPLY (mV)

MAX4249-MAX4257



460

440

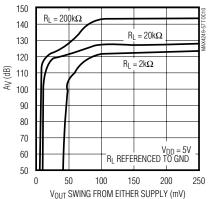
(HA)

SUPPLY CURRENT

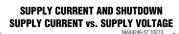
PER AMPLIEIER

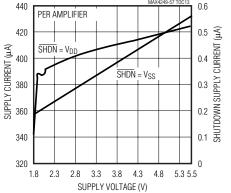
 $(V_{DD} = 5V, V_{SS} = 0, V_{CM} = V_{OUT} = V_{DD}/2$, input noise floor of test equipment =10nV/ \sqrt{Hz} for all distortion measurements, $T_A = +25^{\circ}C$, unless otherwise noted.)

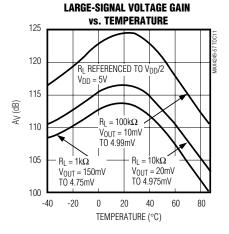
LARGE-SIGNAL VOLTAGE GAIN vs. OUTPUT VOLTAGE SWING



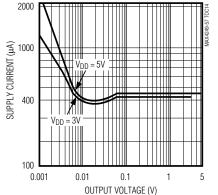
MAX4249-MAX4257











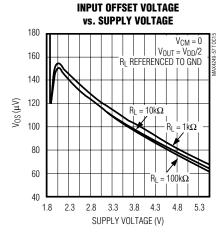
420 0.375 400 $\overline{SHDN} = V_{DD}$ 380 374 SHDN = VSS 360 340 0.373 -20 20 40 60 80 -40 0 TEMPERATURE (°C)

SUPPLY CURRENT AND SHUTDOWN

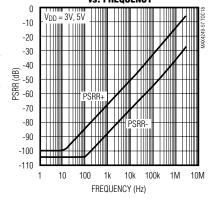
SUPPLY CURRENT vs. TEMPERATURE

0.376

SHUTDOWN SUPPLY CURRENT (µA)



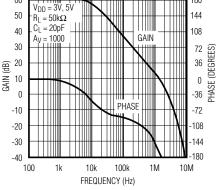
MAX4250-MAX4254 POWER-SUPPLY REJECTION RATIO vs. FREQUENCY



MAX4250-MAX4254 GAIN AND PHASE vs. FREQUENCY 60 180 V_{DD} = 3V, 5V 50 144 $= 50 k\Omega$ R $C_I = 20 pF$ 40 108 = 1000 Av 30 72 GAIN PHASE (DEGREES) 20 36 (gB) GAIN (0 10 0 -36 -10 72 PHASE 108 -20 -30 144 -40 -180 100 1k 10k 100k 1M 10M FREQUENCY (Hz)

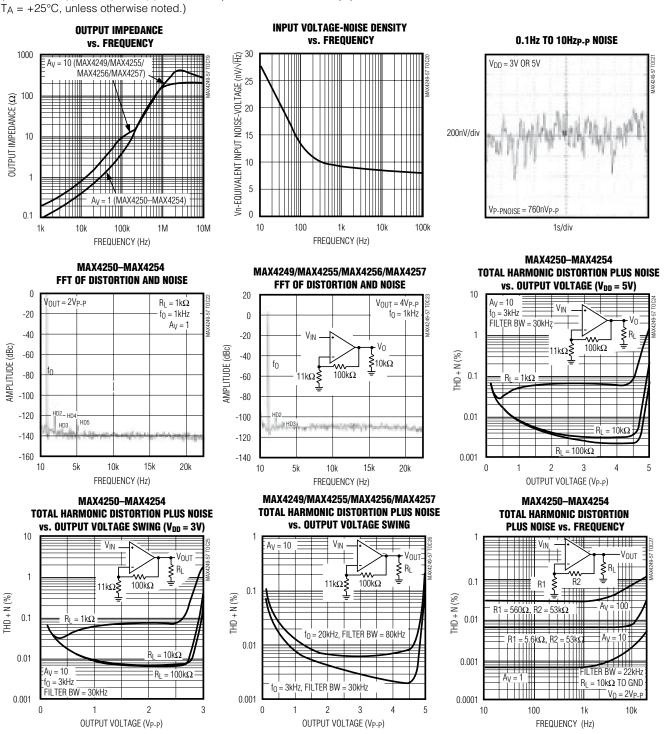
MAX4249/MAX4255/MAX4256/MAX4257 GAIN AND PHASE vs. FREQUENCY MAX4249-57 TOCI7

180



M/IXI/M

Typical Operating Characteristics (continued)

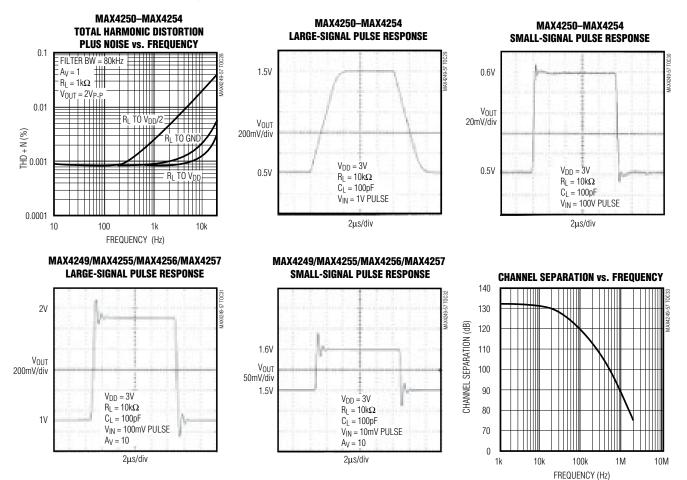


(V_{DD} = 5V, V_{SS} = 0, V_{CM} = V_{OUT} = V_{DD}/2, input noise floor of test equipment =10nV/√Hz for all distortion measurements,

MAX4249-MAX4257

_Typical Operating Characteristics (continued)

 $(V_{DD} = 5V, V_{SS} = 0, V_{CM} = V_{OUT} = V_{DD}/2$, input noise floor of test equipment =10nV/ \sqrt{Hz} for all distortion measurements, $T_A = +25^{\circ}C$, unless otherwise noted.)



M/IXI/M

_Pin Description

			PIN/BUM	P					
MAX4250/ MAX4255	MAX4251/ MAX4256	MAX4252/ MAX4257	MAX4252		MAX4249/ MAX4253		MAX4254	NAME	FUNCTION
5-Pin SOT23	8-Pin SO/µMAX	8-Pin SO/µMAX	8-Pin UCSP	10-Pin UCSP	10-Pin μMAX	14-Pin SO	14-Pin SO		
1	6	1, 7	A1, A3	A1, C1	1, 9	1, 13	1, 7, 8, 14	OUT, OUTA, OUTB, OUTC, OUTD	Amplifier Output
2	4	4	C2	B4	4	4	11	V _{SS}	Negative Supply. Connect to ground for single- supply operation
3	3	3, 5	C1, C3	A3, C3	3, 5	3, 11	3, 5, 10, 12	IN+, INA+, INB+, INC+, IND+	Noninverting Amplifier Input
4	2	2, 6	B1, B3	A2, C2	2, 6	2, 12	2, 6, 9, 13	IN-, INA-, INB-, INC-, IND-	Inverting Amplifier Input
5	7	8	A2	B1	8	14	4	V _{DD}	Positive Supply
_	8	_	_	A4, C4	_	5, 9	_	SHDN, SHDNA, SHDNB	Shutdown Input, Connect to V _{DD} or leave unconnected for normal operation (amplifier(s) enabled).
_	1, 5	_			_	5, 7, 8, 10		N.C.	No Connection. Not internally connected.
_		_	B2	B2, B3				_	Not populated with solder sphere

Detailed Description

The MAX4249–MAX4257 single-supply operational amplifiers feature ultra-low noise and distortion while consuming very little power. Their low distortion and low noise make them ideal for use as preamplifiers in wide dynamic-range applications, such as 16-bit analog-to-digital converters (see *Typical Operating Circuit*). Their high-input impedance and low noise are also useful for signal conditioning of high-impedance sources, such as piezoelectric transducers.

These devices have true rail-to-rail ouput operation, drive loads as low as ${\rm 1k}\Omega$ while maintining DC accura-

cy, and can drive capactive loads up to 400pF without oscillation. The input common-mode voltage range extends from V_{DD} - 1.1V to 200mV beyond the negative rail. The push-pull output stage maintains excellent DC characteristics, while delivering up to ±5mA of current.

The MAX4250–4254 are unity-gain stable, whereas, the MAX4249/MAX4255/MAX4256/MAX4257 have a higher slew rate and are stable for gains \geq 10V/V. The MAX4249/MAX4251/MAX4253/MAX4256 feature a low-power shutdown mode, which reduces the supply current to 0.5µA and disables the outputs.



MAX4249-MAX4257

Low Distortion

Many factors can affect the noise and distortion that the device contributes to the input signal. The following guidelines offer valuable information on the impact of design choices on Total Harmonic Distortion (THD).

Choosing proper feedback and gain resistor values for a particular application can be a very important factor in reducing THD. In general, the smaller the closedloop gain, the smaller the THD generated, especially when driving heavy resistive loads. Large-value feedback resistors can significantly improve distortion. The THD of the part normally increases at approximately 20dB per decade, as a function of frequency. Operating the device near or above the full-power bandwidth significantly degrades distortion.

Referencing the load to either supply also improves the part's distortion performance, because only one of the MOSFETs of the push-pull output stage drives the output. Referencing the load to midsupply increases the part's distortion for a given load and feedback setting. (See the Total Harmonic Distortion vs. Frequency graph in the *Typical Operating Characteristics.*)

For gains \geq 10V/V, the decompensated devices MAX4249/MAX4255/MAX4256/MAX4257 deliver the best distortion performance, since they have a higher slew rate and provide a higher amount of loop gain for a given closed-loop gain setting. Capacitive loads below 400pF, do not significantly affect distortion results. Distortion performance remains relatively constant over supply voltages.

Low Noise

The amplifier's input-referred, noise-voltage density is dominated by flicker noise at lower frequencies, and by thermal noise at higher frequencies. Because the thermal noise contribution is affected by the parallel combination of the feedback resistive network (RF II RG, Figure 1), these resistors should be reduced in cases where the system bandwidth is large and thermal noise is dominant. This noise contribution factor decreases, however, with increasing gain settings.

For example, the input noise-voltage density of the circuit with RF = 100k Ω , RG = 11k Ω (A_V = 10V/V) is e_n = 15nV/ \sqrt{Hz} , e_n can be reduced to 9nV/ \sqrt{Hz} by choosing RF = 10k Ω , RG = 1.1k Ω (A_V = 10V/V), at the expense of greater current consumption and potentially higher distortion. For a gain of 100V/V with RF = 100k Ω , RG = 1.1k Ω , the e_n is low (9nV/ \sqrt{Hz}).

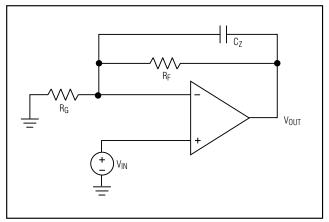
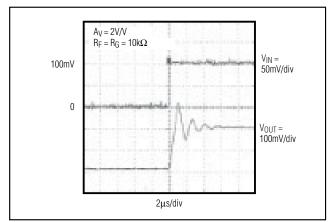
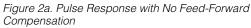


Figure 1. Adding Feed-Forward Compensation





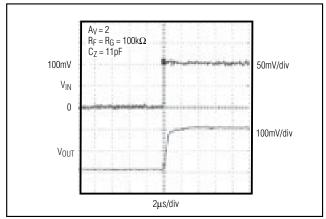


Figure 2b. Pulse Response with 10pF Feed-Forward Compensation



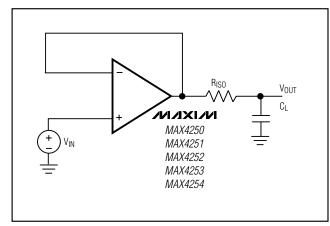


Figure 3. Overdriven Input Showing No Phase Reversal

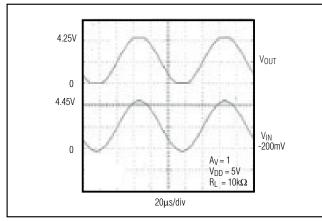


Figure 4. Rail-to-Rail Output Operation

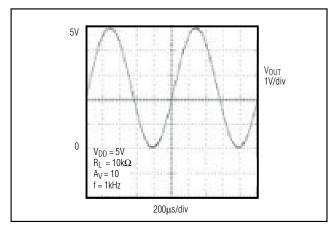


Figure 5. Capacitive-Load Driving Circuit

Using a Feed-Forward Compensation Capacitor, Cz

The amplifier's input capacitance is 11pF. If the resistance seen by the inverting input is large (feedback network), this can introduce a pole within the amplifier's bandwidth, resulting in reduced phase margin. Compensate the reduced phase margin by introducing a feed-forward capacitor (C_Z) between the inverting input and the output (Figure 1). This effectively cancels the pole from the inverting input of the amplifier. Choose the value of C_Z as follows:

$$C_{Z} = 11 \times (R_{F} / R_{G}) [pF]$$

In the unity-gain stable MAX4250–MAX4254, the use of a proper C_Z is most important for A_V = 2V/V, and A_V = -1V/V. In the decompensated MAX4249/MAX4255 /MAX4256/MAX4257, C_Z is most important for A_V = 10V/V. Figures 2a and 2b show transient response both with and without C_Z.

Using a slightly smaller C_Z than suggested by the formula above achieves a higher bandwidth at the expense of reduced phase and gain margin. As a general guideline, consider using C_Z for cases where R_G II R_F is greater than 20k Ω (MAX4250–MAX4254) or greater than 5k Ω (MAX4249/MAX4255/MAX4256/MAX4257).

Applications Information

The MAX4249–MAX4257 combine good driving capability with ground-sensing input and rail-to-rail output operation. With their low distortion, low noise and lowpower consumption, these devices are ideal for use in portable instrumentation systems and other low-power, noise-sensitive applications.

Ground-Sensing and Rail-to-Rail Outputs

The common-mode input range of these devices extends below ground, and offers excellent commonmode rejection. These devices are guaranteed not to undergo phase reversal when the input is overdriven (Figure 3).

Figure 4 showcases the true rail-to-rail output operation of the amplifier, configured with $A_V = 10V/V$. The output swings to within 8mV of the supplies with a $10k\Omega$ load, making the devices ideal in low-supply-voltage applications.

Output Loading and Stability

Even with their low quiescent current of 400μ A, these amplifiers can drive $1k\Omega$ loads while maintaining excellent DC accuracy. Stability while driving heavy capacitive loads is another key feature.

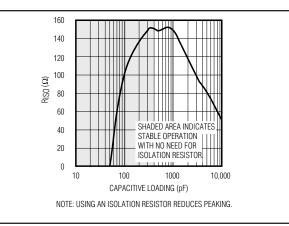


Figure 6. Isolation Resistance vs. Capacitive Loading to Minimize Peaking (<2dB)

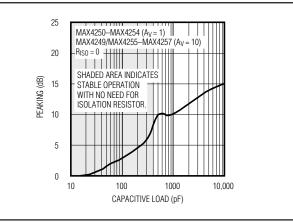


Figure 7. Peaking vs. Capacitive Load

These devices maintain stability while driving loads up to 400pF. To drive higher capacitive loads, place a small isolation resistor in series between the output of the amplifier and the capacitive load (Figure 5). This resistor improves the amplifier's phase margin by isolating the capacitor from the op amp's output. Reference Figure 6 to select a resistance value that will ensure a load capacitance that limits peaking to <2dB (25%). For example, if the capacitive load is 1000pF, the corresponding isolation resistor is 150 Ω . Figure 7 shows that peaking occurs without the isolation resistor. Figure 8 shows the unity-gain bandwidth vs. capacitive load for the MAX4250–MAX4254.

Power Supplies and Layout

The MAX4249–MAX4257 operate from a single 2.4V to 5.5V power supply or from dual supplies of $\pm 1.20V$ to $\pm 2.75V$. For single-supply operation, bypass the power

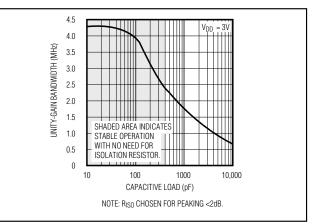


Figure 8. MAX4250-4254 Unity-Gain Bandwidth vs. Capacitive Load

supply with a $0.1\mu F$ ceramic capacitor placed close to the V_{DD} pin. If operating from dual supplies, bypass each supply to ground.

Good layout improves performance by decreasing the amount of stray capacitance and noise at the op amp's inputs and output. To decrease stray capacitance, minimize PC board trace lengths and resistor leads, and place external components close to the op amp's pins.

UCSP Package Consideration

For general UCSP package information and PC layout considerations, please refer to the Maxim Application Note (Wafer-Level Ultra-Chip-Board-Scale-Package).

UCSP Reliability

The UCSP represents a unique packaging form factor that may not perform equally to a packaged product through traditional mechanical reliability tests. UCSP reliability is integrally linked to the user's assembly methods, circuit board material, and usage environment. The user should closely review these areas when considering use of a UCSP. Performance through operating life test and moisture resistance remains uncompromised as it is primarily determined by the wafer-fabrication process. Mechanical stress performance is a greater consideration for a UCSP. UCSPs are attached through direct solder contact to the user's PC board, foregoing the inherent stress relief of a packaged product lead frame. Solder-joint contact integrity must be considered. Table 1 shows the testing done to characterize the UCSP reliability performance. In conclusion, the UCSP is capable of performing reliably through environmental stresses as indicated by the results in the table. Additional usage data and recommendations are detailed in the UCSP application note, which can be found on Maxim's website at www.maxim-ic.com.



_Typical Operating Circuit

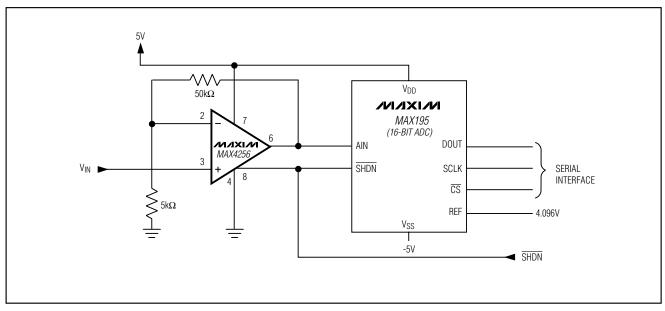


Table 1. Reliability Test Data

TEST	CONDITIONS	DURATION	NO. OF FAILURES PER SAMPLE SIZE
Temperature Cycle	-35°C to +85°C, -40°C to +100°C	150 cycles, 900 cycles	0/10, 0/200
Operating Life	$T_A = +70^{\circ}C$	240h	0/10
Moisture Resistance	-20°C to +60°C, 90% RH	240h	0/10
Low-Temperature Storage	-20°C	240h	0/10
Low-Temperature Operational	-10°C	24h	0/10
Solderability	8h steam age	—	0/15
ESD	±2000V, Human Body Model		0/5
High-Temperature Operating Life	T _J = +150°C	168h	0/45

Selector Guide

PART	GAIN BANDWIDTH (MHz)	MINIMUM STABLE GAIN (V/V)	NO. OF AMPLIFIERS PER PACKAGE	SHUTDOWN MODE	PIN-PACKAGE
MAX4249	22	10	2	Yes	10-pin μMAX, 14-pin SO
MAX4250	3	1	1	—	5-pin SOT23
MAX4251	3	1	1	Yes	8-pin μMAX/SO
MAX4252	3	1	2	—	8-pin μMAX/SO, 8-pin UCSP
MAX4253	3	1	2	Yes	10-pin μMAX, 14-pin SO, 10-pin UCSP
MAX4254	3	1	4	_	14-pin SO
MAX4255	22	10	1	—	5-pin SOT23
MAX4256	22	10	1	Yes	8-pin μMAX/SO
MAX4257	22	10	2	—	8-pin μMAX/SO

_Ordering Information (continued)

PART	TEMP RANGE	PIN- PACKAGE	TOP MARK
MAX4251ESA	-40°C to +85°C	8 SO	_
MAX4251EUA	-40°C to +85°C	8μΜΑΧ	—
MAX4252EBL-T*	-40°C to +85°C	8 UCSP-8	AAO
MAX4252ESA	-40°C to +85°C	8 SO	_
MAX4252EUA	-40°C to +85°C	8 µMAX	—
MAX4253EBC-T*	-40°C to +85°C	10 UCSP-10	AAK
MAX4253EUB	-40°C to +85°C	10 µMAX	—
MAX4253ESD	-40°C to +85°C	14 SO	—
MAX4254ESD	-40°C to +85°C	14 SO	—
MAX4255EUK-T	-40°C to +85°C	5 SOT23-5	ACCJ
MAX4256ESA	-40°C to +85°C	8 SO	—
MAX4256EUA	-40°C to +85°C	8 µMAX	—
MAX4257ESA	-40°C to +85°C	8 SO	—
MAX4257EUA	-40°C to +85°C	8 µMAX	—

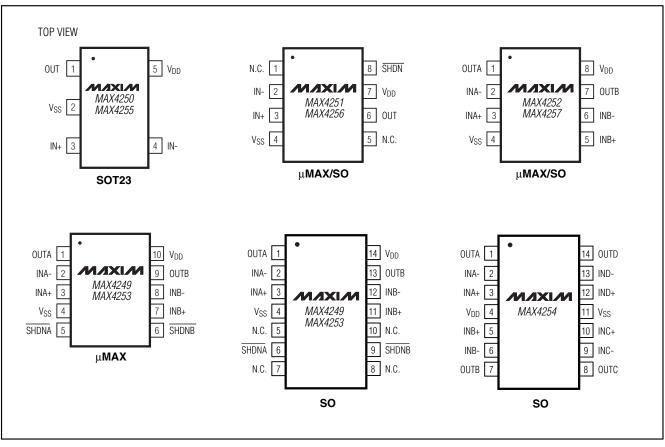
*UCSP reliability is integrally linked to the user's assembly methods, circuit board material, and environment. Refer to the UCSP Reliability Notice in the UCSP Reliability section of this data sheet for more information.

Chip Information

MAX4250/MAX4251/MAX4255/MAX4256 TRANSISTOR COUNT: 170

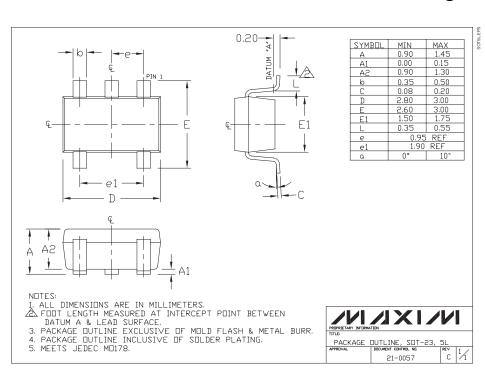
MAX4249/MAX4252/MAX4253/MAX4257 TRANSISTOR COUNT: 340

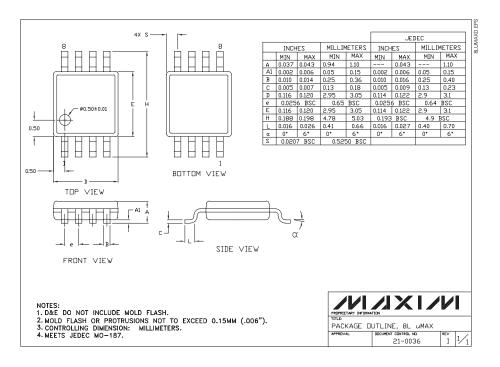
MAX4254 TRANSISTOR COUNT: 680



Pin Configurations (continued)

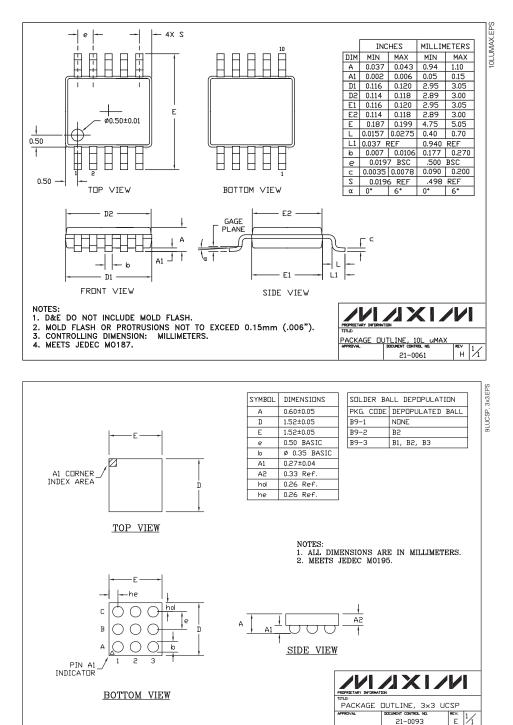
Package Information





M/IXI/M

Package Information (continued)



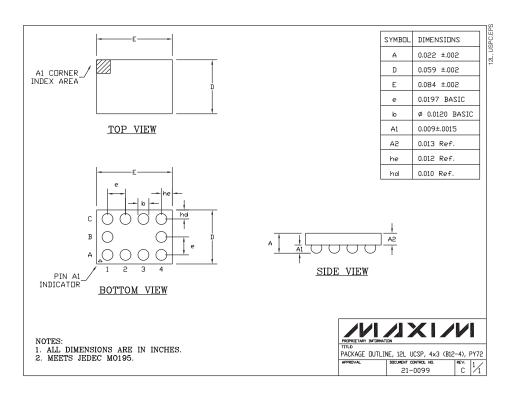
ΕН Ħ E E E 0°-8° Ħ А INCHES MILLIMETERS INCHES MILLIMETERS MIN MAX MIN MAX N MS012 MIN MAX MIN MAX Α 0.053 0.069 1.35 1.75 D 0.189 0.197 4.80 5.00 8 Α A1 0.004 0.010 0.10 0.25 D 0.337 0.344 8.55 8.75 14 В B 0.014 0.019 0.35 0.49 D 0.386 0.394 9.80 10.00 16 С 0.007 0.010 С 0.19 0.25 е 0.050 1.27 NDTES: 1. D&E DO NOT INCLUDE MOLD FLASH 2. MOLD FLASH OR PROTRUSIONS NOT TO EXCEED 15mm (.006*) 0.150 0.157 3.80 4.00 Ε
 H
 0.228
 0.244
 5.80
 6.20

 h
 0.010
 0.020
 0.25
 0.50

 L
 0.016
 0.050
 0.40
 1.27
 3. LEADS TO BE COPLANAR WITHIN .102mm (.004") 4. CONTROLLING DIMENSION: MILLIMETER 5. MEETS JEDEC MS012-XX AS SHOWN IN ABOVE TABLE 6. N = NUMBER OF PINS PACKAGE FAMILY DUTLINE: SDIC .150" 1 21-0041 A

Package Information (continued)

Package Information (continued)



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